

Supplementary Information for:

## A Ferroelectric Semiconductor Field-Effect Transistor

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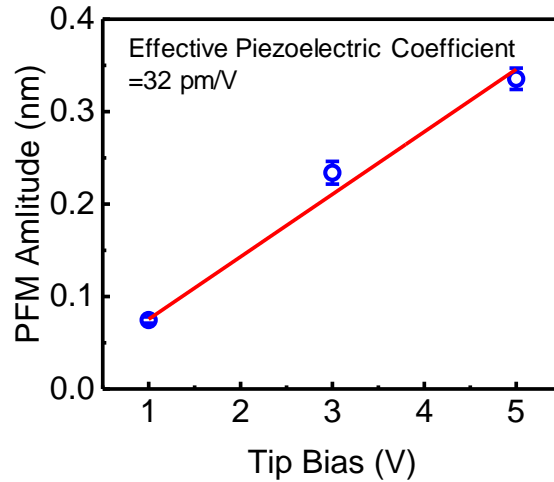
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## 1. Piezoelectric coefficient and DART-PFM data

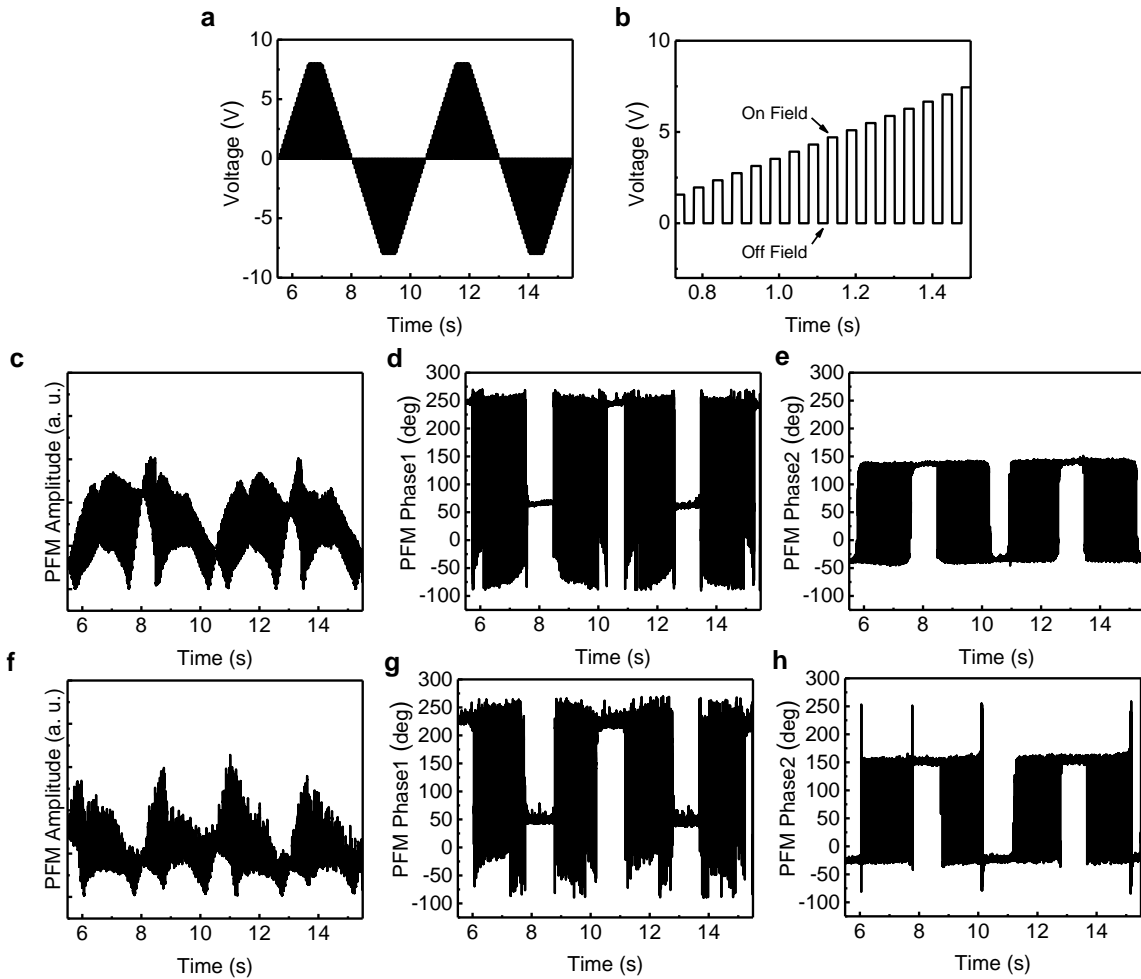


**Figure S1.** Effective piezoelectric coefficient of a 78.7 nm thick  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> flake, measured by PFM.

To extract the piezoelectric coefficient, different AC voltages are applied on the sample from the conductive AFM tip which shows a linear relationship between the mechanical deformation (PFM amplitude) and the electric field, as shown in Fig. S1. The piezoelectric coefficient ( $d_{33}$ ) of the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> flake is 32 pm/V. It should be noticed that the  $d_{33}$  obtained here is effective piezoelectric coefficient ( $d_{33,eff}$ ), which is affected by other tensor elements from the sample and tip-sample electrostatic interaction.

Fig. S2 shows the raw data of single point DART-PFM hysteresis loop measurement measured at room temperature. Fig. S2a shows the applied voltage biases versus time. Two cycles of triangular voltage waves are applied with both on field and off field PFM measurements, as shown in Fig. S2b. Fig. S2c-e show the PFM amplitude, phase1 and phase2 signals, measured using MSM structure. Fig. S2f-h show the PFM amplitude, phase1 and phase2 signals, measured using MOS structure. A clear ferroelectric polarization switching can be seen on both phase1 and

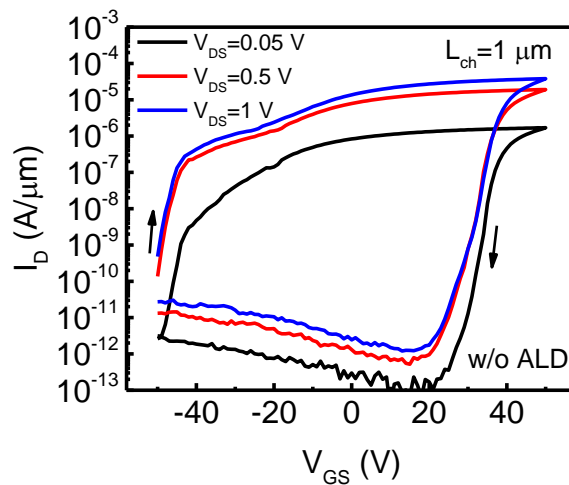
phase2 signals, at both on field and off field and for both MSM and MOS structure. The raw data itself confirms the ferroelectricity and switchable polarization of  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> for FeS-FET application.



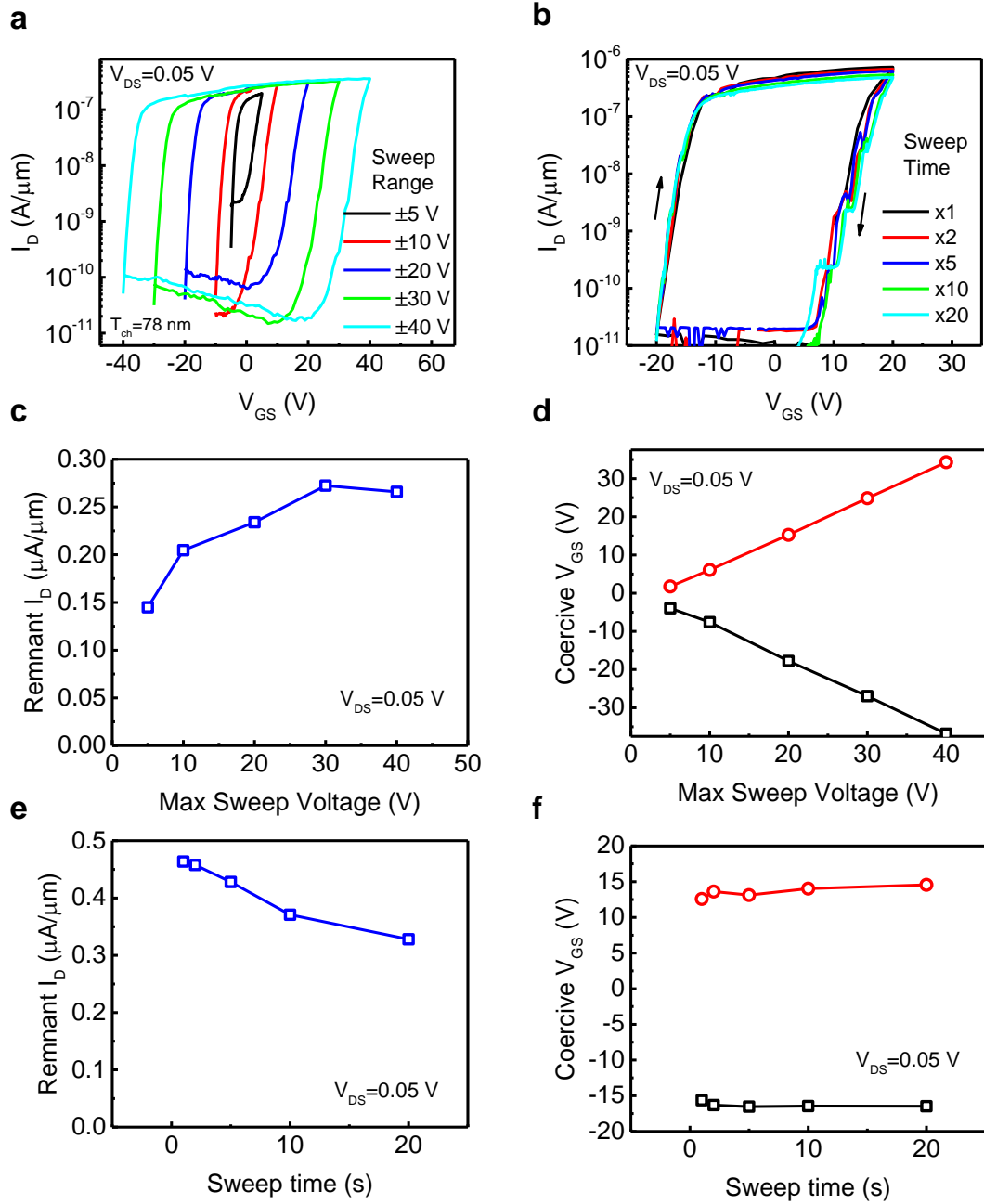
**Figure S2.** Raw data of single point DART-PFM hysteresis loop measurements at room temperature. **a**, Bias versus time and **b**, zoom-in plot of **a**, showing on field and off field measurements. **c**, Amplitude, **d**, phase1, and **e**, phase2 results in MSM structure. **f**, Amplitude, **g**, phase1, and **h**, phase2 results in MOS structure.

## 2. Unpassivated $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs

Fig. S3 shows the  $I_D$ - $V_{GS}$  characteristics of a representative  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FET without ALD passivation, measured by double gate voltage sweep and at different  $V_{DS}$ . The device has a channel length ( $L_{ch}$ ) of 1  $\mu\text{m}$ , channel thickness ( $T_{ch}$ ) of 62.2 nm. The transfer curve shows clear clockwise hysteresis loop and a large memory window over 70 V. A high on/off ratio over  $10^7$  at  $V_{DS}=0.5$  V between on- and off-states is also achieved. Fig. S4 investigates the impact of gate voltage sweep range and sweep time on the performance of  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs and are measured on a  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FET with  $L_{ch}=1$   $\mu\text{m}$  and  $T_{ch}=78$  nm. Fig. S4a and Fig. S4b show the  $I_D$ - $V_{GS}$  characteristics measured at different  $V_{GS}$  sweep ranges and different sweep time (sweep time controlled by the number of  $V_{GS}$  step, the fastest measurement time of the whole loop is about 1 s). Fig. S4c and Fig. S4d plot the remnant  $I_D$  and coercive  $V_{GS}$  versus gate voltage sweep range. The gate voltage sweep range dependence suggests more polarization charge is generated by higher gate voltage. Fig. S4e and Fig. S4f show the remnant  $I_D$  and coercive  $V_{GS}$  versus sweep time. The  $I_D$ - $V_{GS}$  curve has a weak sweep speed dependence, indicating the mobile trapped charges play a minor role in device characteristics.



**Figure S3.**  $I_D$ - $V_{GS}$  characteristics of an  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FET without ALD passivation, measured at  $V_{DS}=0.05$  V and different  $V_{GS}$  sweep ranges. The device has a channel length ( $L_{ch}$ ) of 1  $\mu\text{m}$ , channel thickness ( $T_{ch}$ ) of 62.2 nm.

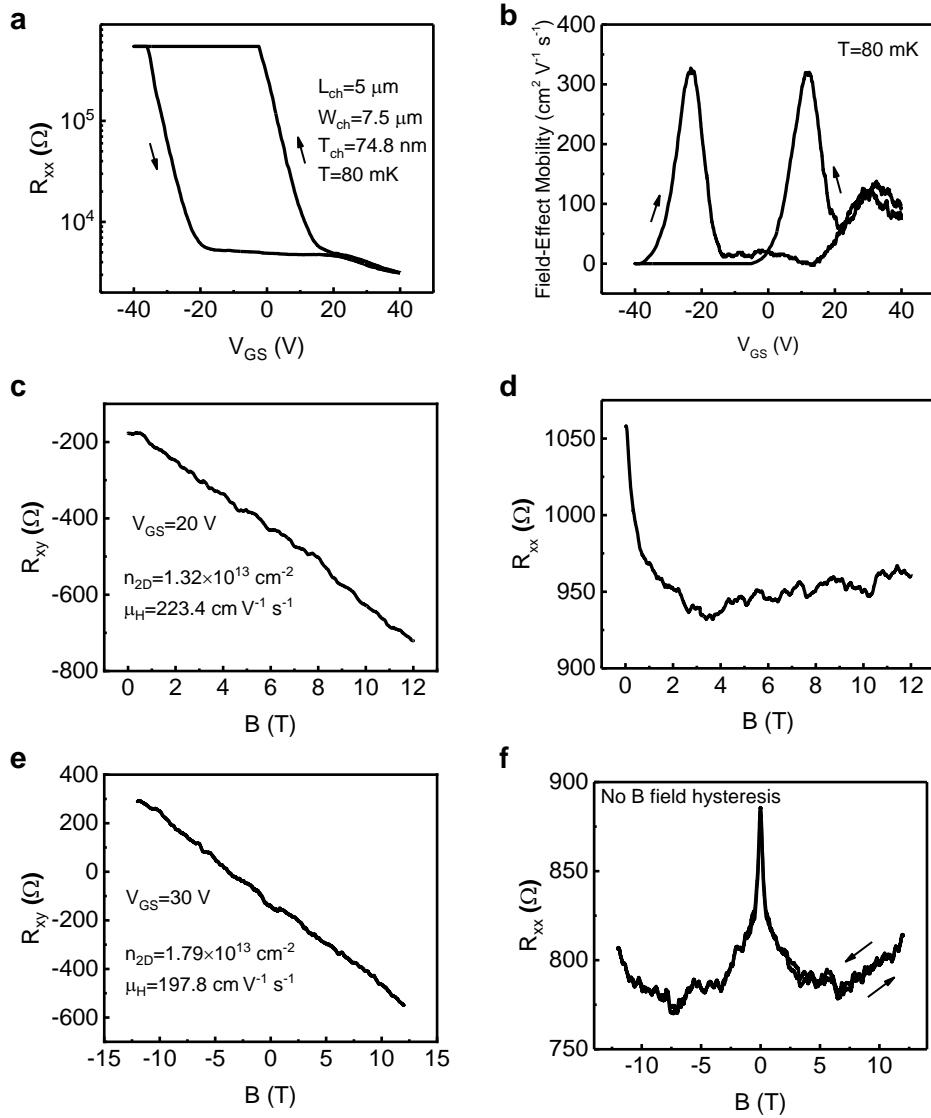


**Figure S4.** **a**,  $I_D$ - $V_{GS}$  characteristics of an  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FET without ALD passivation, measured at  $V_{DS}=0.05$  V and different  $V_{GS}$  sweep ranges. **b**,  $I_D$ - $V_{GS}$  characteristics of the same  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FET, measured at  $V_{DS}=0.05$  V and different sweep times. **c**, Remnant  $I_D$  measured at different  $V_{GS}$  sweep ranges on the same device. **d**, Coercive  $V_{GS}$  ( $V_{GS}$  at  $I_D=10$  nA/ $\mu$ m) measured at different  $V_{GS}$  sweep ranges on the same device. **e**, Remnant  $I_D$  measured at different  $V_{GS}$  sweep times on the same device. **f**, Coercive  $V_{GS}$  ( $V_{GS}$  at  $I_D=10$  nA/ $\mu$ m) measured at different  $V_{GS}$  sweep times on the same device.

### 3. Low temperature and Hall measurements

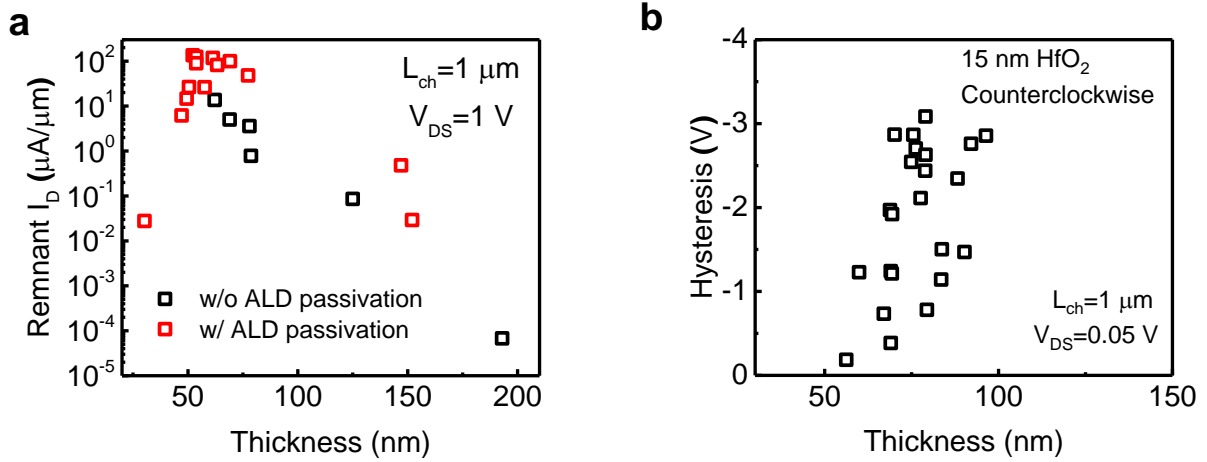
In the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs with 90 nm SiO<sub>2</sub> as gate insulator, the hysteresis loop is clockwise. The hysteresis direction is same as the hysteresis direction induced by charge trapping. To distinguish from charge trapping induced hysteresis loop and prove the ferroelectricity, low temperature I-V and Hall measurement are performed at 80 mK. The clockwise hysteresis still exists at very low temperature measurement down to 80 mK and the hysteresis window is very similar to the room temperature results, as shown in Fig. S5a and S5b. It is well-known that charge trapping process is related with thermal activation and has strong temperature dependence. The threshold voltage shift and transistor hysteresis are reduced significantly at low temperature<sup>1,2</sup>. Therefore, the existence of similar hysteresis window at low temperature of 80 mK supports the existence of ferroelectricity and ferroelectric polarization switching as the origin of the hysteresis.

Fig. S5c and S5d show Hall measurements of  $R_{xx}$  and  $R_{xy}$  versus magnetic field (B) measured at  $V_{GS}=20$  V (same device as Fig. S5a). The corresponding carrier density ( $n_{2D}$ ) and Hall mobility ( $\mu_H$ ) are determined to be  $n_{2D}=1.32\times 10^{13}$  cm<sup>-2</sup> and  $\mu_H=223.4$  cm<sup>2</sup> V<sup>-1</sup> s<sup>-1</sup>. Fig. S5e and S5f show some additional Hall measurement data measured at  $V_{GS}=30$  V and with B field measured bi-directionally (forward and reverse), suggesting no ferromagnetic property exists in this material.



**Figure S5.** **a**,  $R_{xx}$ - $V_{GS}$  characteristics of a  $\alpha$ - $\text{In}_2\text{Se}_3$  FeS-FET with 90 nm  $\text{SiO}_2$  as gate dielectric, 10 nm  $\text{Al}_2\text{O}_3$  capping and using two terminal setup. The device has a channel length of 5  $\mu\text{m}$ , channel width of 7.5  $\mu\text{m}$  and channel thickness of 74.8 nm. All data in this figure are measured on the same device and at very low temperature of 80 mK. **b**, Field-effect mobility calculated from **a**. **c**,  $R_{xy}$  versus B field characteristics measured at  $V_{GS}=20$  V, showing a 2D electron density of  $1.32 \times 10^{13} \text{ cm}^{-2}$  and a hall mobility of  $223.4 \text{ cm}^2 \text{ V}^{-1} \text{ s}^{-1}$ . **d**,  $R_{xx}$  versus B field characteristics at  $V_{GS}=20$  V, showing a negative magnetoresistance with quantum oscillations due to low carrier mobility. **e**,  $R_{xy}$  versus B field characteristics measured at  $V_{GS}=30$  V, showing a 2D electron density of  $1.79 \times 10^{13} \text{ cm}^{-2}$  and a Hall mobility of  $197.8 \text{ cm}^2 \text{ V}^{-1} \text{ s}^{-1}$ . **f**,  $R_{xx}$  versus B field characteristics measured bi-directionally at  $V_{GS}=30$  V, showing a weak-localization-like negative magnetoresistance and no obvious B field hysteresis.

#### 4. Channel thickness dependence in $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs



**Figure S6. a**, Channel thickness dependence and comparison of remnant drain current (at  $V_{\text{GS}}=0 \text{ V}$  and  $V_{\text{DS}}=1 \text{ V}$  in forward sweep) versus channel thickness of  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs (90 nm SiO<sub>2</sub>) with and without ALD Al<sub>2</sub>O<sub>3</sub> passivation. Significant on-current improvement is achieved by ALD Al<sub>2</sub>O<sub>3</sub> passivation. **b**, Hysteresis versus thickness for  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs with 15 nm HfO<sub>2</sub> as gate insulator and with 10 nm ALD Al<sub>2</sub>O<sub>3</sub> passivation.

Fig. S6a shows the channel thickness dependent remnant  $I_D$  of the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs with  $I_D$ - $V_{\text{GS}}$  curve measured at  $V_{\text{DS}}=1 \text{ V}$ . Devices with and without ALD Al<sub>2</sub>O<sub>3</sub> passivation are compared. The remnant  $I_D$  versus channel thickness has a peak position at around 50-70 nm, and decreases exponentially while  $T_{\text{ch}}$  increases or decreases beyond this thickness range. For thicker channel, the reason of the decrease is because the maximum voltage applied (40-50 V) is not sufficiently high to trigger the ferroelectric polarization switching. For thinner channel, this might be due to the ferroelectricity in  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> is getting weaker in thinner channel, but it hasn't been clearly understood yet at current stage. Therefore, the thickness dependence also suggests the ferroelectric polarization is critical to the performance and operation of the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> FeS-FETs. The maximum remnant  $I_D$  at  $V_{\text{DS}}=1 \text{ V}$  in devices with ALD passivation ( $135 \mu\text{A}/\mu\text{m}$ ) is found to be significantly higher (nearly one order of magnitude) than that of devices without ALD passivation ( $13.7 \mu\text{A}/\mu\text{m}$ ). The hysteresis window size is found to have a wide distribution versus



thickness, as shown in Fig. S6b. The strong correlation between channel thickness and switching voltages is not obvious. This is another evidence of the competition between the bottom surface conduction and the top surface conduction. If the surface Fermi level pinning and surface passivation of the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> Fe-FET can be further engineered and improved, the uniformity of the hysteresis window must be improved.

## 5. Discussions on the accuracy of field-effect mobility

The field effect mobility ( $\mu_{FE}$ ) is defined as<sup>3</sup>,

$$\mu_{FE} = \frac{Lg_m}{WC_{ox}V_{DS}} \quad (1)$$

$$g_m = \frac{dI_D}{dV_{GS}} \quad (2)$$

The transconductance ( $g_m$ ) is usually extracted from the transfer ( $I_D$ - $V_{GS}$ ) characteristics of the devices. It is no doubt that the  $\mu_{FE}$  can always be calculated from eqns. (1) and (2). However, the accuracy of the  $\mu_{FE}$  need further discussion because  $\mu_{FE}$  may not be accurate to measure the intrinsic mobility ( $\mu$ ) of the material. But it can reflect transport property of the material and the related device. There are two possible origins for the inaccurate estimation.

The first is because of the gate voltage dependence of intrinsic mobility<sup>4</sup>. The conductance is  $\sigma = n_{ch}q\mu$ , and channel carrier density  $n_{ch} = C_{ox}(V_G - V_T)/q$ . Thus, without considering the contact resistance,  $\mu_{FE}$  can be re-write as,

$$\mu_{FE} = \frac{1}{C_{ox}} \frac{d\sigma}{dV_G} \quad (3)$$

$$\mu_{FE} = \mu + (V_G - V_T) \frac{d\mu}{dV_G} \quad (4)$$

From eqn. (4),  $\mu_{FE}$  is under-estimated if  $\mu$  become smaller at higher  $V_G$ .

The second origin is the Schottky contact resistance ( $R_C$ ) (this time  $\mu$  is assumed to be a constant for simplicity). So, we have the conductance and  $\mu_{FE}$  re-write as,

$$\sigma = \frac{1}{R_C + \frac{1}{n_{ch}q\mu}} \quad (5)$$

$$\mu_{FE} = \frac{1}{(1 + n_{ch}q\mu R_C)^2} (\mu - n_{ch}q\mu(V_G - V_T) \frac{dR_C}{dV_G}) \quad (6)$$

If  $R_C$  is a constant,  $\mu_{FE}$  is under-estimated as shown in eqn. (7).

$$\mu_{FE} = \frac{\mu}{(1 + n_{ch}q\mu R_C)^2} \quad (7)$$

If  $R_C$  is not a constant and becomes smaller at higher  $V_G$ ,  $\mu_{FE}$  can be either over- or under-estimated depends on the  $R_C$  properties.

## 6. Theory and simulation of FeS-FET

Device level simulations have been conducted to investigate the clockwise and counterclockwise hysteresis in the I-V characteristics of FeS-FET. More specifically, we have performed physics based self-consistent simulation of FeS-FET devices by coupling Poisson's equation, Ginzburg-Landau equation and 2D charge equation. For simplicity, a 1D cut from the source to gate region (along  $z$ -axis) of FeS-FET device has been adopted in our simulation framework. A van der Waals gap is assumed between the source/drain contacts and the semiconductor channel because of the 2D layered nature of  $\alpha$ -In<sub>2</sub>Se<sub>3</sub>. Now, according to the Ginzburg-Landau theory, an equation for polarization and electric field can be written as,

$$E_{FE} + K(\nabla^2 P) = \alpha P + \beta P^3 \quad (1)$$

Here,  $P$  is the polarization,  $E_{FE}$  is the electric-field and  $K$  is the domain coupling coefficient of the ferroelectric semiconductor and all of them are defined for out-of-plane direction. Also,  $\alpha$  and  $\beta$  are the effective Landau coefficients that describe the dielectric stiffness of the ferroelectric semiconductor material. The P-E relation by assumption is shown in Fig. S7. Now, the total displacement ( $D$ ) in each layer can be defined as,

$$D = \epsilon_0 \epsilon_r E_{FE} + P$$

Here,  $\epsilon_0$  and  $\epsilon_r$  are the vacuum and background permittivity, respectively. The Poisson's equation within the ferroelectric semiconductor layer can be written as,

$$-\epsilon_r \frac{\partial^2 \phi}{\partial z^2} = -\frac{\partial P}{\partial z} + \rho_f \quad (2)$$

$$\rho_f = q(-n + p)$$

Here,  $\phi$  is the electrostatic potential,  $\rho_f$  is the mobile charge density.  $n$  and  $p$  are the electron density and hole density, respectively and have been calculated by using the following equations.

$$n = \frac{g_C}{t_l} \times kT \times \log \left( 1 + \exp \left( \frac{E_F - E_C}{kT} \right) \right); \quad g_C = \frac{2m_C^*}{\pi \hbar^2} \quad (3)$$

$$p = \frac{g_V}{t_l} \times kT \times \log \left( 1 + \exp \left( \frac{E_V - E_F}{kT} \right) \right); \quad g_V = \frac{2m_V^*}{\pi \hbar^2} \quad (4)$$

Here,  $g_C$  and  $g_V$  are the 2D density of states of electron and hole, respectively.  $m_C^*$  and  $m_V^*$  are the effective mass of electron and hole, respectively.  $E_C$ ,  $E_V$  and  $E_F$  are the conduction-band energy, valence-band energy and fermi-level energy, respectively. Also,  $k$  is the Boltzmann constant,  $\hbar$  is the reduced Plank's constant,  $t_l$  is the thickness of each  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> layer and  $T$  is the temperature. In simulation, source electro-chemical potential has been assumed to be equal to the channel fermi-level. Finally, to get the electrostatic device characteristics we solve equation (1)-(4) self consistently to calculate polarization distribution, charge and potential profile.

From the self-consistent electrostatic simulation, we get the value of electron and hole charge density. By assuming the electron is the dominant contributor in the conduction mechanism, a drift current has been approximated by using the following equation.

$$I'_{DS} = [Q'_e + 1 \times 10^{-7} Q'_h] \times \mu \times \frac{V_{DS}}{L_{CH}}$$

$$Q_{e(h)'} = \sum_{i=1}^n q \times t_l n(p)_i$$

Here,  $I'_{DS}$  is the current per unit area.  $Q_{e(h)}'$  is the channel electrons (holes) per unit area, respectively, which have been calculated by summing up the mobile electron (hole) charges in all the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> layers.

The coupled ferroelectric and semiconducting nature of  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> is critical to analyze for understanding the device operation of FeS-FET. Each of the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> layers can exhibit either a

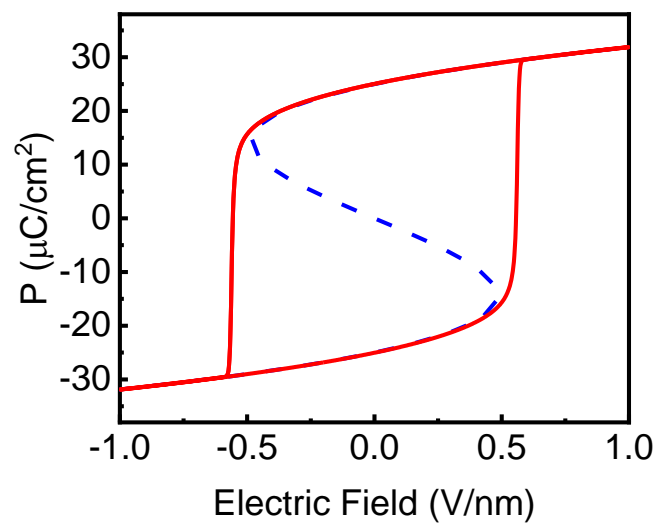
positive or negative polarization and that polarization can be switched by applying a gate voltage. However, the extent of polarization switching depends on the amount of electric field in the  $\alpha$ - $\text{In}_2\text{Se}_3$  channel and that further is dependent on the gate dielectric thickness for an applied gate voltage. To that effect, we consider two different EOT: high EOT (30 nm) and low EOT (0.5 nm) to analyze the device operations.

- High EOT devices:

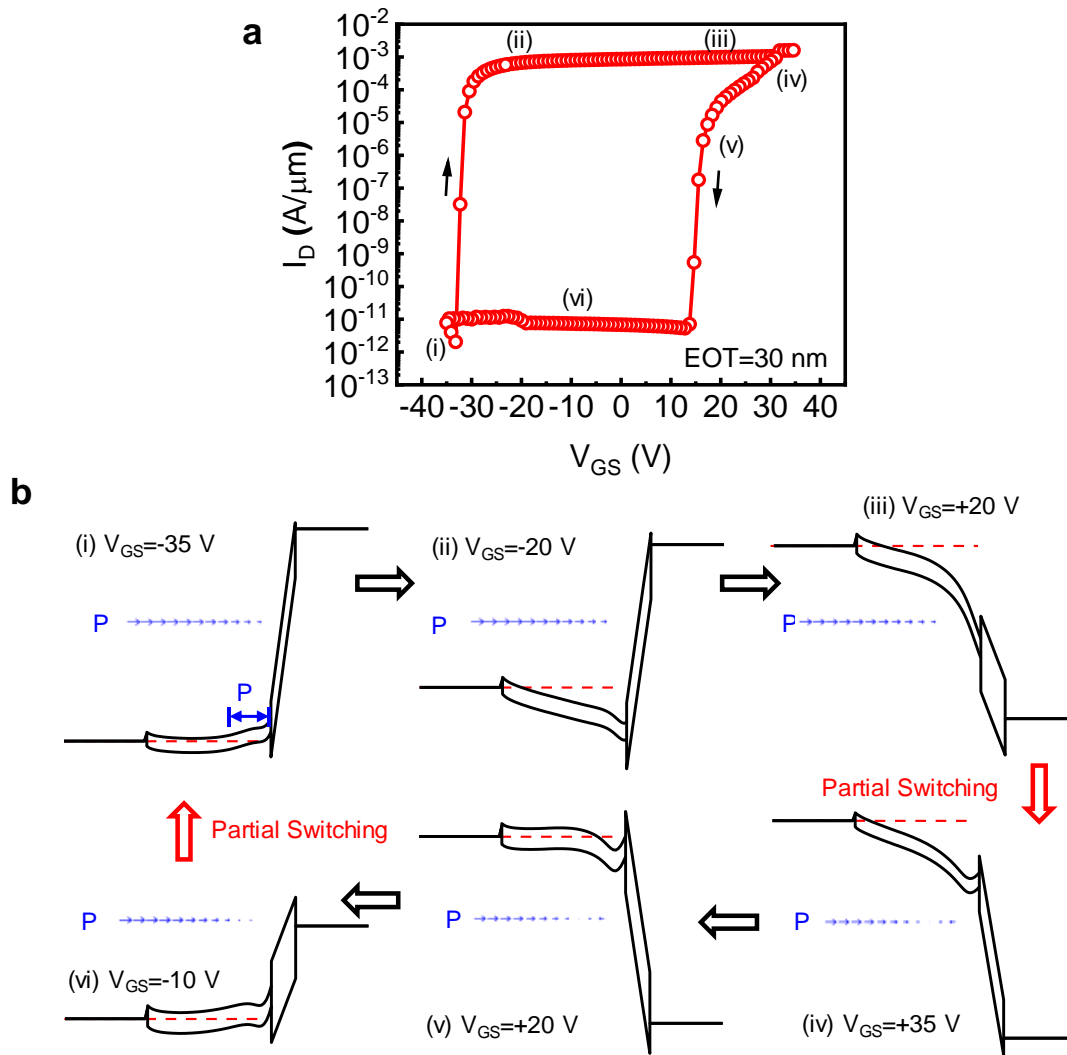
First, let us assume that the applied gate voltage is highly negative (-35 V) and therefore, all the  $\alpha$ - $\text{In}_2\text{Se}_3$  layers exhibit polarization down (as shown in Fig. S8b(i)). Now, with the increase in gate voltage, polarization starts to change and the bottom  $\alpha$ - $\text{In}_2\text{Se}_3$  layer (at gate oxide/semiconductor interface) conduction band goes below the fermi level (Fig. S8b(i)-(iii)). Consequentially, electron density increases near the bottom  $\alpha$ - $\text{In}_2\text{Se}_3$  layer and that gives rise to an increase in channel current as shown in Fig. S8b(i)-(iii). At such scenario, bottom  $\alpha$ - $\text{In}_2\text{Se}_3$  layers are the most dominant participant in channel current. Interestingly, with a further increase in gate voltage, the polarization starts switching (from down to up) from the bottom layer of  $\alpha$ - $\text{In}_2\text{Se}_3$  (Fig. S8b(iv)). At this point, the channel exhibits both up polarization (in first or first few bottom layers) and down polarization (in rest of the top layers) and that gives rise to a domain-wall (DW) within the channel. Note that, after the down to up polarization switching in bottom few  $\alpha$ - $\text{In}_2\text{Se}_3$  layers, the minima of conduction band edge shifts from the bottom most layer to the vicinity of the DW. We refer such situation as the ‘partially polarization switching’ and in our considered case such phenomena happen at the gate voltage of  $\sim 30\text{V}$ . Now, with the decrease in gate voltage, the conduction band moves upward and consequently, current decreases rapidly. Such operation leads to clockwise hysteresis in  $I_D$ - $V_{GS}$  characteristics.

- Low EOT devices:

Similar to the high EOT devices, with the increase in gate voltage channel current also increases in low EOT devices (as shown in Fig. S9b(i)-(iii)). Once the bottom layer polarization is switched at a certain positive gate voltage, further increase in gate voltage will induce a layer by layer gradual polarization switching. At the same time, the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> layer corresponding to the DW moves further from the bottom to the top  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> layer. At a sufficiently high and positive gate voltage, all the layer will switch to polarization up and the corresponding band diagram is being depicted in Fig. S9b(iv)-(v). In such condition, both the top and bottom layers are conducting. Now, with the decrease in gate voltage, the conduction band edge near the bottom layers will start going above the fermi-level and therefore, the electron density contribution from bottom layer will decrease (Fig. S9b(v)-(vi)). However, the top layer is still conducting. Hence, a significant amount of current can be observed even at a negative gate voltage (-3.24V). However, at a sufficiently high negative voltage, all the  $\alpha$ -In<sub>2</sub>Se<sub>3</sub> layers will switch back to down polarization and consequently, the top layers will be no more conducting. Therefore, an abrupt decrease in current can be observed (as shown in Fig. S9b(vi)-(i)). Such operation leads to a counterclockwise hysteresis in  $I_{DS}$ - $V_{GS}$  characteristics of FeS-FET.

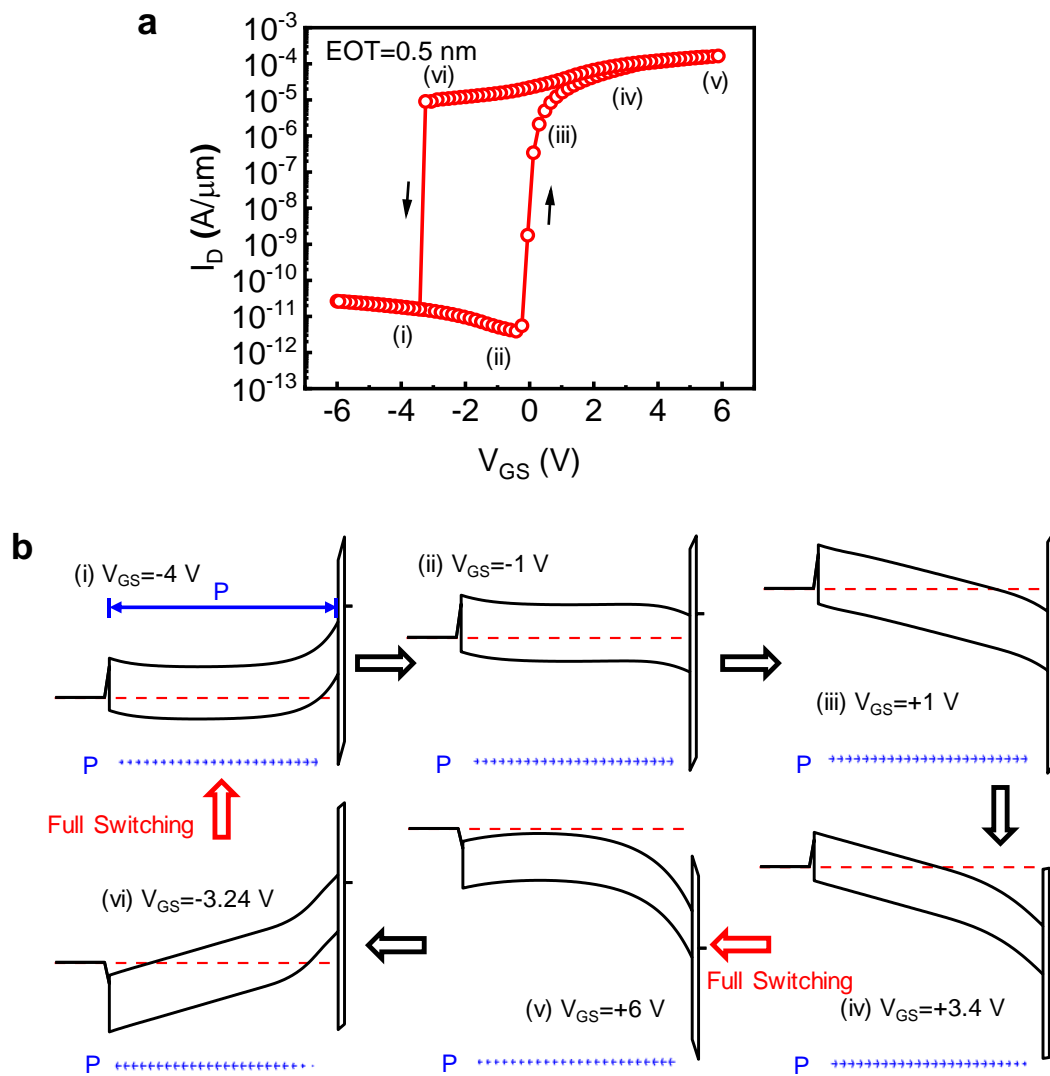


**Figure S7.** Polarization versus electric field characteristics by assumption in the simulation.



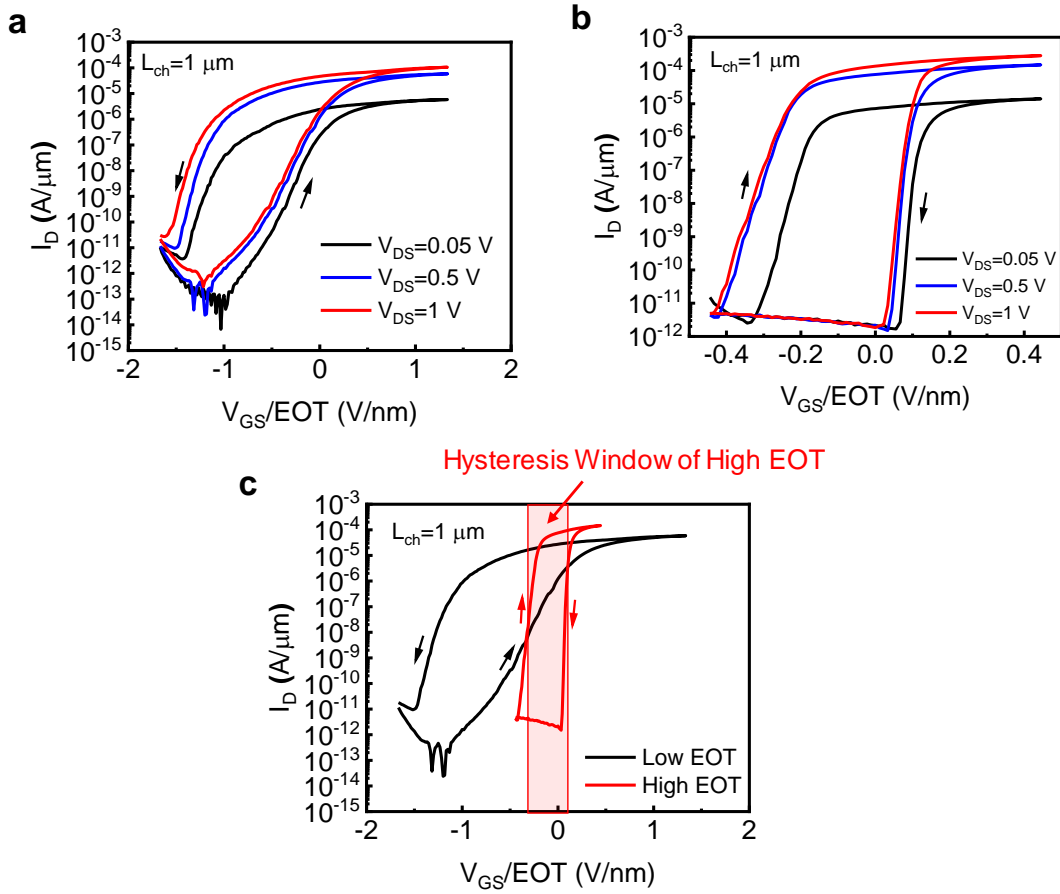
**Figure S8.** **a**, Simulation of  $I_D$ - $V_{GS}$  characteristics of  $\alpha$ - $\text{In}_2\text{Se}_3$  FeS-FET at EOT=30 nm. **b**, Band diagram and polarization vector map at different gate voltages during bi-directional gate voltage sweep. The direction of polarization vectors near insulator/semiconductor interface in (iv) and (v) is opposite to those in bulk and top surface.





**Figure S9.** **a**, Simulation of  $I_D$ - $V_{GS}$  characteristics of  $\alpha$ - $\text{In}_2\text{Se}_3$  FeS-FET at EOT=0.5 nm. **b**, Band diagram and polarization vector map at different gate voltages during bi-directional gate voltage sweep.

## 7. Partial Switching of FeS-FET in Low EOT Condition

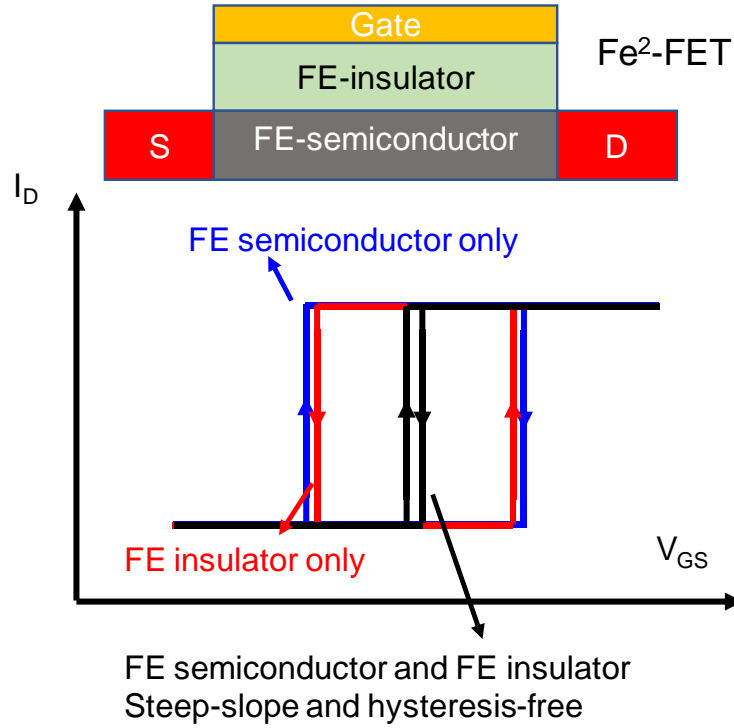


**Figure S10.** **a**,  $I_D$  as a function of  $V_{GS}/EOT$  for FeS-FET with low EOT. **b**,  $I_D$  as a function of  $V_{GS}/EOT$  for FeS-FET with high EOT. **c**, Comparison of  $I_D$  as a function of  $V_{GS}/EOT$  between low EOT and high EOT at  $V_{DS}=0.5\ \text{V}$ .

To compare the electrical performance of  $\alpha\text{-In}_2\text{Se}_3$  FeS-FETs with different EOT under similar displacement field in gate oxide, we re-plot the transfer characteristics by introducing new x-axis as  $V_{GS}/EOT$ . Note that as displacement field is continuous (if no mobile charge), same displacement field in oxide is equivalent to same electric field in semiconductor. Fig. S10a shows  $I_D$  as a function of  $V_{GS}/EOT$  for a FeS-FET with low EOT and Fig. S10b shows  $I_D$  as a function of  $V_{GS}/EOT$  for a FeS-FET with high EOT. The slope in the subthreshold region is equal to  $SS/EOT$  by definition. Fig. S10c shows the comparison of  $I_D$  as a function of  $V_{GS}/EOT$  between low EOT and high EOT at  $V_{DS}=0.5\ \text{V}$ . It is clear that device with high EOT shows a much sharper

slope ( $SS/EOT$ ) in subthreshold region in this new  $I_D$  versus  $V_{GS}/EOT$  curve.  $SS/EOT$  is much smaller at high EOT condition. A direct result is that in the hysteresis window of high EOT condition (partial switching condition), as illustrated in Fig. S10c, the device with low EOT is either at off-state in the subthreshold region or at on-state. Therefore, if applying this partial switching condition to the device with low EOT, it cannot be effectively turn-on and turn-off. In addition, to change the electric field direction inside the semiconductor, on-state to off-state or off-state to on-state transitions are necessary (to change the direction of band bending). The change of applied electric field direction is the necessary condition of ferroelectric polarization switching in the ferroelectric semiconductor. In the on-state or off-state at low EOT condition, by applying same  $V_{GS}/EOT$  window, the electric field in the semiconductor doesn't even change the direction (Fig. S10(c)), so a partial switching induced clockwise hysteresis loop is difficult to achieve for low EOT devices.

## 8. A proposal of a deep steep-slope and hysteresis-free all ferroelectric Fe<sup>2</sup>-FET



**Figure S11.** Illustration of a deep steep-slope and hysteresis-free ferroelectric insulator and ferroelectric semiconductor all ferroelectric field-effect transistors (Fe<sup>2</sup>-FET).

As another step further by using the counterclockwise hysteresis of a Fe-FET and clockwise hysteresis of a FeS-FET, we propose to integrate a ferroelectric insulator on a ferroelectric semiconductor channel, as shown in Fig. S11 as a Fe<sup>2</sup>-FET, to eliminate the hysteresis loops in both Fe-FET and FeS-FET and achieve a new type of deep steep-slope and hysteresis-free transistor. The key point is to match the two ferroelectric polarization charge so that the net ferroelectric polarization charge on the FE-insulator/FE-semiconductor interface is close to zero, so that hysteresis-free operation can be achieved.

## References

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